

Neural-network-based nonlinear iterative learning control: Magnetic brake study

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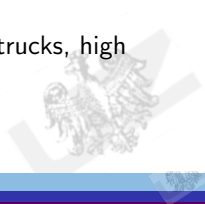


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Introduction

- Intelligent control – a very popular and important branch of control methods
- Iterative learning control (ILC) – modern intelligent control strategy
- Neural networks – useful when dealing with nonlinear problems
 - nonlinear plant modeling
 - time-varying nonlinear learning controller realization
- Robustness of ILC – to consider state and output disturbances/uncertainty as well as initial state and initial feedback controller errors
- Experimental verification – magnetic brake system
- Magnetic brakes are often used in different areas such as big trucks, high speed railways, commercial vehicles or industrial elevators



Problem formulation

Consider a class of discrete-time nonlinear systems

$$\begin{aligned}\mathbf{x}_p(k+1) &= g(\mathbf{x}_p(k), u_p(k)) + \mathbf{w}_p(k), \quad k = 0, \dots, N-1, \\ y_p(k) &= \mathbf{C}\mathbf{x}_p(k) + v_p(k),\end{aligned}\tag{1}$$

where $p \geq 0$ – a trial number, N – a trial length

$\mathbf{x}_p(k)$, $u_p(k)$, $y_p(k)$ – system state, input and response

$\mathbf{w}_p(k)$, $v_p(k)$ – state and output disturbances/uncertainty

g – some nonlinear function

\mathbf{C} – output (observation) matrix

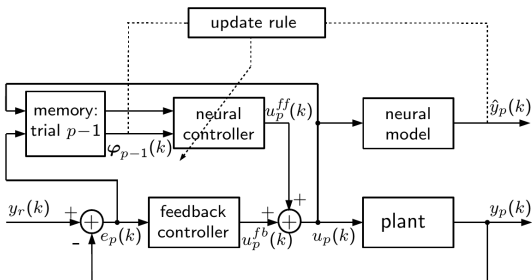
Problem: to design a control scheme such that the tracking error $e_p(k) = y_r(k) - y_p(k)$ ($y_r(k)$ – the reference trajectory) remains bounded in the presence of model uncertainty and disturbances

Iterative learning control

- Adaptation of so-called first-order ILC scheme (*current iteration ILC*)
 - use of existing feedback controller for stabilization,
 - adding supporting feedforward **neural** controller for tracking improvement,

$$u_p(k) = u_p^{fb}(k) + u_p^{ff}(k) \quad (2)$$

where p – trial number, $u_p^{fb}(k)$ – feedback control, $u_p^{ff}(k)$ – ILC update



- Feedback controller

$$\begin{aligned}z_p(k+1) &= \mathbf{A}_c z_p(k) + \mathbf{B}_c e_p(k) \\ u_p^{fb}(k) &= \mathbf{C}_c z_p(k) + \mathbf{D}_c e_p(k)\end{aligned}\quad (3)$$

where $z_p(k)$ – the controller state

$\mathbf{A}_c, \mathbf{B}_c, \mathbf{C}_c, \mathbf{D}_c$ – controller matrices

- Learning controller

$$u_p^{ff}(k) = f(u_{p-1}(k), e_{p-1}(k)) \quad (4)$$

where f – a nonlinear function

Basic concept – use the neural network to provide time-varying realization of the function $u_p^{ff}(k)$ (being implicitly an inverted model of the plant)



Neural controller

- Neural network with one hidden layer

$$u_p^{ff}(k) = f(\varphi_{p-1}(k)) = \mathbf{W}_{2,p}\sigma(\mathbf{W}_{1,p}\varphi_{p-1}(k) + \mathbf{b}_{1,p}) + \mathbf{b}_{2,p},$$

where $\varphi_{p-1}(k) = [u_{p-1}(k) \quad e_{p-1}(k)]$

$\mathbf{W}_{1,p}$, $\mathbf{W}_{2,p}$ – weight matrices

$\mathbf{b}_{1,p}$, $\mathbf{b}_{2,p}$ – bias vectors

σ – hidden neurons activation function

- Neural network parameters are updated after each process trial
- Stochastic gradient based training algorithm

K. Patan and M. Patan, "Neural-network-based iterative learning control of nonlinear systems," *ISA Transactions*, vol. 98, pp. 445–453, 2020.



Neural model

- System modeling – state space neural network model

$$\begin{aligned}\hat{\mathbf{x}}_p(k+1) &= \hat{g}(\hat{\mathbf{x}}_p(k), u_p(k), \varepsilon_p(k)) \\ \hat{y}_p(k) &= \mathbf{C}\hat{\mathbf{x}}_p(k)\end{aligned}$$

where $\hat{\mathbf{x}}_p \in \mathbb{R}^n$, $u_p \in \mathbb{R}^1$, $\hat{y}_p \in \mathbb{R}^1$ – model state, input and output
 $\varepsilon_p(k) = y_p(k) - \hat{y}_p(k)$ – prediction error

- Implementation of nonlinear function $\hat{g}(\cdot, \cdot, \cdot)$:

$$\hat{g}(\cdot, \cdot, \cdot) = \mathbf{A}\hat{\mathbf{x}}_p(k) + \mathbf{V}_2\sigma(\mathbf{V}_1^x\hat{\mathbf{x}}_p(k) + \mathbf{V}_1^u u_p(k) + \mathbf{V}_1^\varepsilon \varepsilon_p(k) + \mathbf{V}_1^b) + \mathbf{V}_2^b$$

where $\mathbf{V}_1^u \in \mathbb{R}^{v_m \times 1}$, $\mathbf{V}_1^\varepsilon \in \mathbb{R}^{v_m \times 1}$, $\mathbf{V}_1^x \in \mathbb{R}^{v_m \times n}$, $\mathbf{V}_2 \in \mathbb{R}^{n \times v_m}$ – weight matrices
 $\mathbf{V}_1^b \in \mathbb{R}^{v_m}$, $\mathbf{V}_2^b \in \mathbb{R}^n$ – bias vectors
 $\sigma: \mathbb{R}^{v_m} \rightarrow \mathbb{R}^{v_m}$ – the vector-valued activation function
 v_m – the number of hidden neurons



$$\mathbf{A} = \begin{bmatrix} 0 & \dots & 0 & 0 \\ 1 & \dots & 0 & 0 \\ \vdots & \ddots & \vdots & \vdots \\ 0 & \dots & 1 & 0 \end{bmatrix}, \quad \mathbf{C} = [1 \quad 0 \quad \dots \quad 0],$$

$$\mathbf{V}_1^x = \left[\begin{array}{c} \text{tunable weights} \\ \hline \mathbf{0}_{(v_m-n) \times n} \end{array} \right],$$

$$\mathbf{V}_2 = \left[\begin{array}{c|c} \mathbf{0}_{(n-1) \times n} & \text{tunable weights} \\ \hline \text{tunable weights} & \mathbf{0}_{1 \times (v_m-n)} \end{array} \right].$$

- Training in batch mode (off-line) based on historical measurement data



Convergence analysis

Assumption A1. Let $w_p(k) = 0$, $v_p(k) = 0$ and $y_r(k)$ be a reference defined over a discrete-time $k \in N$, which is assumed to be realizable, that is there exists a unique $u_r(k)$ and an initial state $\mathbf{x}_r(0)$, i.e.

$$\begin{aligned}\mathbf{x}_r(k+1) &= g(\mathbf{x}_r(k), u_r(k)) \\ y_r(k) &= \mathbf{C}\mathbf{x}_r(k)\end{aligned}$$

Assumption A2. Let $\forall k \in N$, $\forall p$ state and output disturbances/uncertainties satisfy

$$\|\mathbf{w}_p(k)\| \leq \epsilon_w, \quad \|v_p(k)\| \leq \epsilon_v,$$

where $\epsilon_w \geq 0$, $\epsilon_v \geq 0$ are finite bounds. Moreover, $\forall p$ the initial system state error and initial feedback controller state satisfy

$$\|\Delta\mathbf{x}_p(0)\| \leq \epsilon_x, \quad \|z_p(0)\| \leq \epsilon_z, \quad (5)$$

where $\Delta\mathbf{x}_p(k) = \mathbf{x}_k(k) - \mathbf{x}_p(k)$, $\epsilon_x \geq 0$ and $\epsilon_z \geq 0$ are some positive constants.

Assumption A3. The nonlinear function g satisfies the global Lipschitz condition

$$\|g(\mathbf{x}_1, u_1) - g(\mathbf{x}_2, u_2)\| \leq L (\|\mathbf{x}_1 - \mathbf{x}_2\| + |u_1 - u_2|)$$

where $L > 0$ stands for the Lipschitz constant.



Theorem 1

Let us consider the nonlinear system (1) which satisfies the assumptions (A1)–(A3) and the reference trajectory $y_r(k)$ satisfying the assumption (A1). Then using the control of the form (2)–(4) satisfying the condition

$$\left| \sup_k \|f_u(k)\| + \sup_k \alpha_k L \frac{1 - \beta^{-(\lambda-1)N}}{\beta^\lambda - \beta} \right| < 1 \quad (6)$$

where $f_u(k) = \frac{\partial f}{\partial u_p^f(k)}$, $f_e(k) = \frac{\partial f}{\partial e_p(k)}$,

$$\alpha_k = \max\{\|f_u(k)\| \|B_c\|, \|f_u(k)\| \|D_c\| \|C\| + \|f_e(k)\| \|C\|\},$$

$$\beta = \max\{L \|C_c\| + \|A_c\|, L + L \|D_c\| \|C\| + \|B_c\| \|C\|\},$$

guarantees that the tracking error is bounded, i.e.

$$\lim_{p \rightarrow \infty} \|y_r(k) - y_p(k)\|_\lambda \leq \sigma \quad (7)$$

where constant $\sigma > 0$ is dependent on $\epsilon_w, \epsilon_v, \epsilon_x, \epsilon_z$.



Sketch of proof

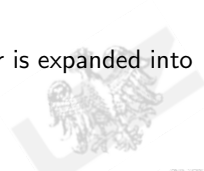
- proof has been obtained as an extension of works:
 1. K. Patan and M. Patan, "Neural-network-based iterative learning control of nonlinear systems," *ISA Transactions*, vol. 98, pp. 445–453, 2020
 2. C.-J. Chien, "A discrete iterative learning control for a class of nonlinear time-varying systems," *IEEE Transactions on Automatic Control*, vol. 43, no. 5, pp. 748–752, 1998
- proof is based on deriving uniform convergence property

$$\lim_{p \rightarrow \infty} u_p(k) = u_r(k),$$

through analysis of the induced norm imposed on the control law

$$\|s(k)\|_\lambda = \sup_{k \in [0, N-1]} \beta^{-\lambda k} \|s(k)\|$$

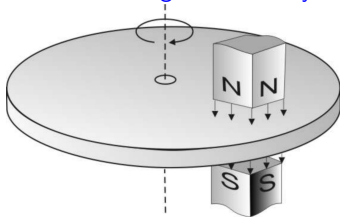
- to deal with a nonlinear representation, the learning controller is expanded into Taylor series
- recursive nature of the state-space representation is also used



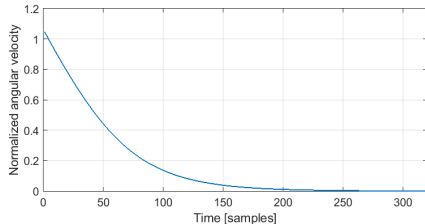
Illustrative example – magnetic brake

- System consists of a magnet inducing currents in a rotating disc of conductive material
- Aluminum disk with a radius of 10cm and a thickness of 1cm
- System input – the magnetic flux, system output – the angular velocity
- Initial value of velocity: 200 RPS

Scheme of magnetic brake system



Reference profile



Model developing

- Magnetic brake is a nonlinear system governed by spatio-temporal dynamics
- Problems with modeling:
 - physical effects such as nonlinear saturation, skin effects and eddy currents induced by motion must be considered simultaneously
 - a fine spatial mesh is required due to very small skin depths
 - a transient solution with time-stepping is necessary
- State-space neural networks provide an important alternative
- Investigated is stable – data recorded in the open-loop control feeding the system with different input signals
- Structure of the state-space neural network model: the model order: 3, the number of hyperbolic tangent neurons of the first layer: 15, the number of linear neurons of the second layer: 3
- Model training: Levenberg-Marquardt method
- Lipschitz constant for the trained model: $L = 1.66$



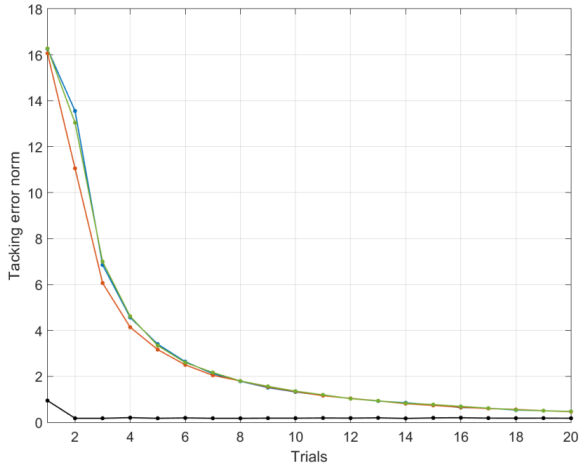
ILC controller synthesis

- open-loop ILC is used (without feedback controller)
- random initial neural controller parameters
- training dataset: $\{e(k), u(k)\}_{k=1}^N$ – recorded during the previous working cycle of the system
- structure of the neural controller: $v_c = 12, \sigma_h \equiv \tanh$
- training carried out after each trial



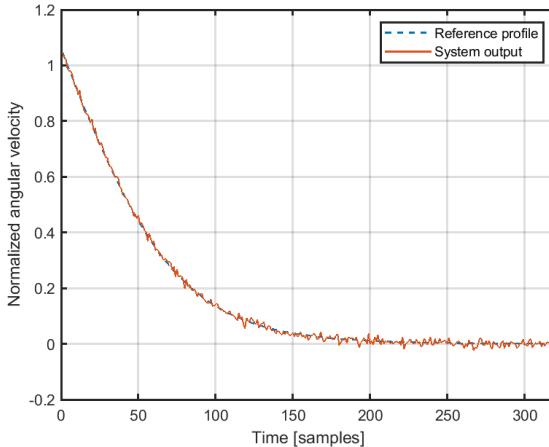
Results – control

Convergence: random initial parameters (blue, red, green)
preliminarily trained controller (black)



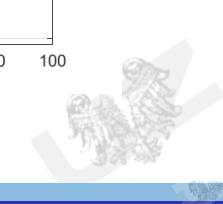
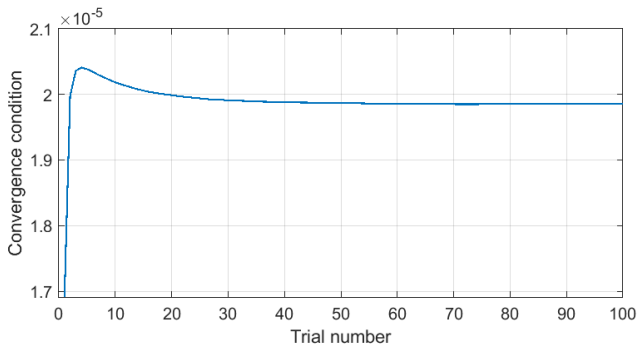
Results – output tracking

Output vs reference after 20 trails



Results – convergence condition satisfaction

Convergence condition for $\lambda = 10$



Concluding remarks

- A novel approach for robust ILC synthesis based on neural networks is proposed
- Learning controller has time-varying structure
- Modeling uncertainty and disturbances are taken into account
- Sufficient conditions guaranteeing convergence of the proposed neural-network-based ILC are provided
- Future research directions:
 - experiments using current-iteration setting
 - comparative studies with alternative control schemes dedicated to magnetic brake system

